

# Lecture 22, Mar 28, 2024

## Frequency Response Design Method

- For a (stable) LTI system  $G(s)$ , the steady-state response to an input  $u(t) = A \sin(\omega_0 t) 1(t)$  is given by  $y_{ss}(t) = A |G(j\omega_0)| \sin(\omega_0 t + \angle G(j\omega_0))$ 
  - The response is a sinusoid of the same frequency, scaled by a factor of  $|G(j\omega_0)|$  (the magnitude of the transfer function, known as the *gain* or *amplitude/magnitude ratio*), with a phase shift of  $\angle G(j\omega_0)$  (the phase of the transfer function)
  - Knowing the magnitude  $M(\omega)$  and phase  $\phi(\omega)$  of  $G(j\omega)$  for all possible frequencies  $\omega$  fully specifies the transfer function
- In general, the complete response is the sum of a number of exponentials and a sinusoid; since the system is stable, all the exponentials decay to 0 as  $t \rightarrow \infty$  and we are only left with the sinusoid
- Example: RC circuit, output  $y(t)$  is the voltage across the capacitor, input  $Ku(t)$  is an input voltage that is sinusoidal
  - $RC \frac{dy}{dt} + y(t) + Ku(t) \implies \frac{dy}{dt} + ky(t) = u(t)$  where  $k = \frac{1}{RC}$ , assuming  $K = RC$
  - $G(s) = \frac{1}{s + k}$
  - Given  $u(t) = \sin(10t) 1(t)$ ,  $U(s) = \frac{10}{s^2 + 100}$
  - At  $s = j10$ ,  $|G(j10)| = \frac{1}{\sqrt{1^2 + 10^2}}$  and  $\angle G(j10) = -\tan^{-1}\left(\frac{10}{1}\right)$
  - Therefore the response is  $y(t) = \frac{1}{\sqrt{101}} \sin(10t - \tan^{-1}(10))$
- Example: lead network  $D_c(s) = K \frac{T_s + 1}{\alpha T_s + 1}$  for  $\alpha < 1$ 
  - Note that this is mathematically identical to the form of the lead compensator we had before, but this form is more common and convenient for frequency response design
    - \* The zero is at  $\frac{1}{T}$ , the pole at  $\frac{1}{\alpha T}$  and the gain is  $\frac{K}{\alpha}$
  - Frequency response:  $D_c(j\omega) = K \frac{T_j \omega + 1}{\alpha T_j \omega + 1}$
  - Gain:  $M = |K| \frac{\sqrt{1 + \omega^2 T^2}}{\sqrt{1 + \alpha^2 \omega^2 T^2}}$
  - Phase:  $\phi = \tan^{-1}(\omega T) - \tan^{-1}(\alpha \omega T)$
  - For  $\omega \rightarrow 0$ , we have  $M \rightarrow |K|$  and  $\phi \rightarrow 0$
  - For  $\omega \rightarrow \infty$  we have  $M \rightarrow \left| \frac{K}{\alpha} \right|$  and  $\phi \rightarrow 0$
- The gain and phase for a range of values of  $\omega$  can be summarized in a *Bode plot*
  - The top plot is the magnitude plot; the bottom plot is the phase plot
  - The bode plot is log-log for magnitude and semi-log for phase
    - \* Using a log-log plot for gain allows us to cover a wide range of  $\omega$  and gain, and also allows us to simply add up the magnitude plots of transfer functions to get the final plot, since multiplication of gains is just addition of logs
  - The vertical axis of the magnitude plot often uses decibels,  $\text{dB} = 20 \log |G(j\omega)|$
- Note in MATLAB, `bode(sys, w)` gives `[mag, phase]`, which we can plot to get the Bode plot
  - Use `logspace()` to get the points for `w`

## System Behaviour From Frequency Response

- The gain and phase of the system's frequency response completely determines the behaviour of the system; we design using it just like we design using the root locus
  - The root locus is to the root locus design method as the Bode plot is to the frequency design method

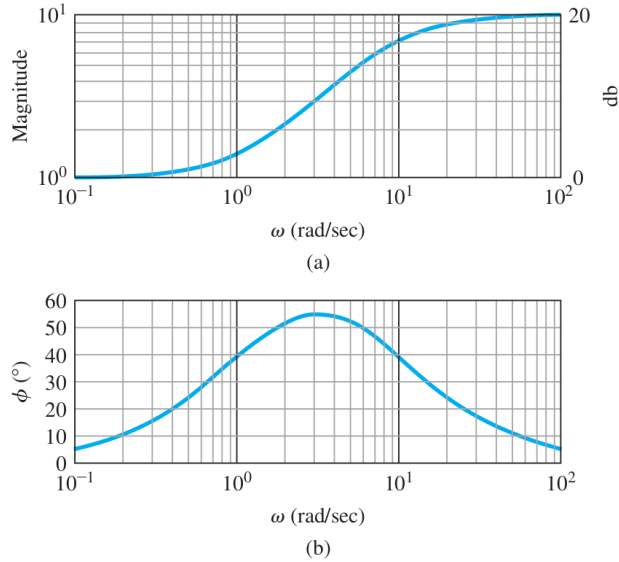


Figure 1: Bode magnitude and phase plots for the lead compensator, for  $K = 1, \alpha = 0.1, T = 1$ .

- Typical closed-loop systems exhibit a low-pass filter behaviour
  - The gain is close to 1 at lower frequencies, i.e. the output follows the input well
  - Beyond a certain frequency, the gain deviates from 1; for most systems, it increases first before decreasing
  - For most systems when the frequency gets very large the gain approaches 0, i.e. the output stops following the input at all
- The *bandwidth*  $\omega_{BW}$  is defined as the highest frequency  $\omega$  where the output still tracks the (sinusoidal) input in a satisfactory manner
  - Traditionally we define this to be when the gain hits  $\sqrt{2}/2 = 0.707$
  - This is known as the *half-power point*; if the gain is a voltage gain, then at this point, the power of the response will be only half
  - A higher bandwidth means a faster response – the larger  $\omega_{BW}$  is, the larger  $\omega_n$  is and the shorter our rise and peak times
- The *resonant peak*  $M_r$  is the maximum value of the amplitude ratio
  - $M_r$  has a direct relationship with  $\zeta$ , so we can estimate the damping and overshoot of the system from  $M_r$
- When we design a controller, we examine its bode plot and tune the gains to get the desired bandwidth and resonant peak, just like we identify pole locations on a root locus

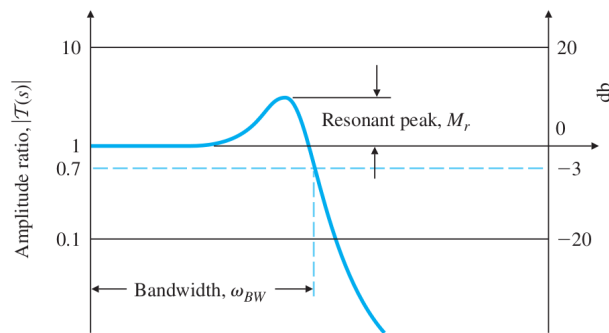


Figure 2: Definitions of bandwidth and resonant peak.